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High-Performance Interior Permanent Magnet Synchronous Motor for Electric Vehicles over Wide Speed Range

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Abstract: Battery electric vehicle (EVs) has become very popular with the interest of environmentally friendly electric vehicles rather than the conventional-Internal Combustion Engine (ICE) vehicles. To run with power, it needs the continuous wide working speed range of motors for the stable power region, strong power in high speed and a great start rallying torque in the low speed. This paper presents a simulation model of an Interior Permanent Magnet Synchronous Motor IPMSM for electric vehicle (EV) traction purpose in MATLAB/Simulink. This consists of Field Oriented Control (FOC), and runs Flux Weakening (FW) controller. 25 The EV market grows rapidly, hence the IPMSM is the best motor for such application due to that the motor has high torque-to-current ratio, high power mathematical eqns. We were able to identify behavior of the motor using FOC control also below base speed, when using FW control at rated speeds; the model indicated response was robust [1]. In addition, we will demonstrate greater high-speed torque using a feedforward flux weakening controller method, with constant power. The responses of speed, below and above rated speed will be stated regarding the results of the simulation. This model approach can suggest a rough level of advice on how to select an appropriate IPMSM configuration regarding distinct vehicle applications and drive conditions.

Index Terms: pure electric vehicle, flux weakening (FW), modelling interior permanent magnet synchronous motor (IPMSM), Field Oriented Control (FOC), FTP75 Drive Cycle.

I. INTRODUCTION

The shift toward cleaner and more sustainable transportation systems is accelerating taking place around the world. The need to address climate change, reduce urban air pollution, and lessen the reliance on dwindling fossil fuel resources all contribute to this development. Within this frame, electric vehicles (EVs) have risen to prominence as a game-changing solution due to their positive environmental effects, energy efficiency, and greater economic sustainability. In contrast to the conventional Internal Combustion Engine (ICE) vehicles which burn hydrocarbon fuels and emit devastatingly carbon-intensive gases EVs support a greener paradigm of sustainable mobility as they rely on electricity stored in onboard batteries. The traction motor, the central component of the propulsion system that transforms electrical energy into mechanical drive force, is a vital enabler of EV performance. The selection of a suitable electric motor significantly influences the overall system efficiency, dynamic responsiveness, and driving comfort. Out of the wide range of motor technologies, the interior permanent magnet synchronous motor (IPMSM) has become very popular in the EV industry. These motors offer high power density, high torque-to-current ratio, large speed range, and high thermal and structural durability [2]. The permanent magnets embedded in the rotor yields more saliency and both magnetic and reluctance torque components thus making IPMSM highly suitable for EV drive cycles especially when stressed under urban and highway duties. To build, model, and perform the evaluation of an IPMSM based drive system adapted for the electric vehicle application is the aim of the research. A primary focus on developing advanced control systems, particularly FoC and FW, which are paramount to utilizing IPMSMs [3]. FOC allows for separate management of torque and flux, providing exceptional accuracy and responsive performance, especially at reduced speeds. Conversely, FW broadens the motor's operational speed range past its fundamental limits by reducing the rotor flux, enabling the system to sustain efficient performance at greater vehicle speeds without surpassing voltage or current limitations [4]. To verify the practicality and efficacy of the suggested control framework, simulations are performed using the Federal Test Procedure (FTP-75) driving cycle, a standardized benchmark that replicates actual urban driving with regular stops and varying accelerations. Through the modelling of motor dynamics and the incorporation of control algorithms within a realistic simulation setting, this research not only showcases theoretical efficacy but also confirms practical relevance in real-world scenarios [5].

Ultimately, this research contributes a comprehensive framework for deploying high-performance IPMSM drives in modern EVs. The findings emphasize the pivotal role of precision control in achieving energy-efficient, high-torque, and reliable operation. As the transportation sector moves toward electrification, IPMSMs when paired with intelligent control schemes are poised to become a foundational technology in the pursuit of efficient, eco-friendly, and intelligent mobility solutions.

II. MATHEMATICAL MODELLING OF IPMSM

An accurate mathematical model is vital for the design, analysis, and simulation of high-performance drive systems, such as electric vehicles (EVs), which demand high efficiency and dynamic performance. The Interior Permanent Magnet Synchronous Motor (IPMSM) is an ideal candidate for traction as well as for other motor technologies since it can produce magnetic torque as well as reluctance torque. To fully exploit the potential of IPMSM, a complete dynamic model is needed, which covers the electrical and mechanical behavior, at various operating points.

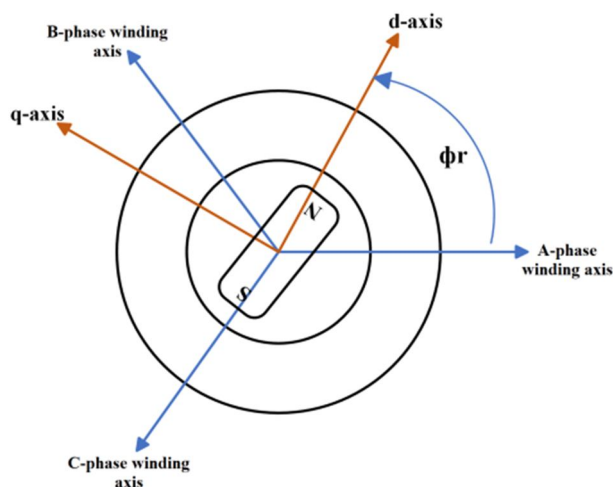


Fig 1: d-q axis reference frame representation

The Fig [1] represents the modeling methodology that converts the original three-phase representation into a two-axis rotating reference frame using the well-known Clarke and Park transformations [6]. This transformation enables the separation of the three phase sinusoidal currents and voltages of the stator into two orthogonal components, known as the direct axis (d-axis) and the quadrature axis (q-axis), which are aligned with the magnetic axis of the rotor. The q-axis is perpendicular to the rotor and generates torque directly while the d-axis is aligned with the rotor permanent magnet flux. By separating the regulation of torque and flux the d-q reference coordinate system reduces regulator complexity and affords the use of modern control mechanisms such as Field Oriented Control (FOC). The eqns [1] & [2] for stator voltage for the IPMSM can be represented as a function of the electrical speed of the motor, time varying flux couples, and resistance of windings in relation to the d-q reference frame. The general forms are:

$$V_d = R_s i_d + \frac{d\lambda_d}{dt} - \omega_e \lambda_q \quad (1)$$

$$V_q = R_s i_q + \frac{d\lambda_q}{dt} + \omega_e \lambda_d \quad (2)$$

where i_d and i_q in eqn [1] & [2] are the stator currents, λ_d and λ_q are the flux linkages in the d- and q-axes, R_s is the stator windings resistance, ω_e is the rotor electrical angular speed, and V_d and V_q are the stator voltages in the d- and q-axes. The flux linkages can be calculated based on the stator inductances and magnetic flux generated by the rotor as given in the eqn [3]:

$$\lambda_d = L_d i_d + \lambda_m \quad (3)$$

In this case, λ_m is the magnetic flux created by the rotor's permanent magnets, and L_d and L_q mentioned in eqn [4] are inductances in the d-axis and q-axis, respectively. The eqn [4] states that the motor also has electromagnet torque, which is the interaction between the stator current and the flux from the permanent magnets, in addition to the reluctance torque induced by the rotor saliency. This torque in eqn [4] can be expressed as follows:

$$T_e = \frac{3}{2} p [\lambda_m i_q + (L_d - L_q) i_d i_q] \quad (4)$$

The term p in eqn [4] is the number of pairs of poles in the motor and T_e is the torque produced by the electro-magnetism. The eqn [5] demonstrates the inherent advantage of the IPMSM over surface mounted PMSM, in that the additional component of reluctance torque can be used to improve performance. The rotor's mechanical behavior is represented through rotational dynamics [7]. The net torque applied to the rotor in eqn [5] causes angular acceleration, countered by mechanical inertia and frictional forces. The relationship is described by:

$$J \frac{d\omega_m}{dt} = T_e - T_L - B\omega_m \quad (5)$$

where T_L is the load torque B is the viscous damping coefficient ω_m the mechanical angular velocity and J is the rotor inertia as shown in eqn [5]. Idealized assumptions are made when constructing operational simulation models that include the aforementioned modes of operation. The main assumptions are that the stator winding is sinusoidally distributed resulting in a balanced magnetic field and that saturation and slotted effects are neglected with constant inductances and linear magnetic behaviour. These simplifications facilitate real-time simulation and control design while preserving essential system dynamics.

The complete IPMSM model, encompassing both electrical and mechanical eqns, can be implemented in a modular form within platforms such as MATLAB/Simulink. This allows for the separation of electrical subsystems governing current, voltage, and flux behavior from the mechanical subsystem, which models rotor dynamics and torque output. This simulation may be used to incorporate control techniques to assess how the motor responds to different driving cycles and load scenarios.

III. FLUX WEAKENING CONTROL STRATEGY FOR HIGH-SPEED IPMSM OPERATION

A. Importance of Voltage and Current Limitations

The IPMSM (interior permanent magnet synchronous motor) utilized in the high-performance EV drive needs to be run properly over as wide of a speed range as possible, extending from zero speed and far beyond its base speed. At the higher speeds, however, the back-emf induced by the rotor's magnetic field rises linearly with rpm. When this back-EMF is close to or exceeds the inverter's DC-link voltage, it is both physically impossible and impractical to control the voltage across the stator winding to your target using standard Field-Oriented Control (FOC) [8]. The suspension will harden to any continued increases in acceleration.

The eqn [6] and [7] states that to prevent the motor drive from violating the voltage and current limits set by the inverter and hardware ratings, it is essential to enforce specific operating boundaries:

$$\sqrt{I_d^2 + I_q^2} \leq I_{max} \quad (6)$$

$$\sqrt{V_d^2 + V_q^2} \leq V_{max} \quad (7)$$

Where V_d and V_q are the voltages, I_d and I_q are the stator current d-axis and q-axis components. The inverter maximum line-to-line voltage (@ sine wave voltage) of a Space Vector PWM (SVPWM) system $V_{max} = \{V_{dc}/\sqrt{3}\}$;

Adhering to these constraints ensures that both the thermal limits of the inverter and the insulation limits of the motor are respected under all operating conditions.

B. Overview of Flux Weakening Control

As the motor transitions into the high-speed region beyond its base speed the flux weakening technique becomes essential for continued speed increase. The essence of flux weakening lies in actively reducing the motor's internal magnetic field to bring down the back-EMF, thus maintaining the required voltage headroom for current regulation. Negative d-axis current is injected into the stator coils, where I_d is acting against the PM flux as opposed to normal operation where the q-axis current is providing torque [9]. The injected negative I_d limits the total air-gap flux linkage so the back-EMF is only limited to a maximum voltage. Some torque capability is reduced because effective flux linking the stator field has been reduced but this is needed to also extend the operating limits of the IPMSM into the constant power region where torque continues to decrease with speed and power is nearly constant.

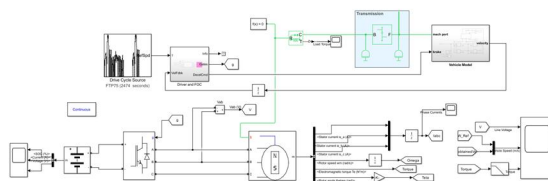
The eqn [8] is the precise calculation of I_d under flux weakening is guided by the following eqn derived from voltage constraint conditions:

$$I_d = \frac{\left(-\Psi_m L_d + \sqrt{(\Psi_m L_d)^2 - (L_d^2 - L_q^2) \left[\Psi_m^2 + L_q^2 I_q^2 - \left(\frac{V_{max}}{\omega} \right)^2 \right]} \right)}{(L_d^2 - L_q^2)} \quad (8)$$

- Ψ_m = flux linkage for permanent magnets
- L_d and L_q = The inductances of the d- and q-axes
- I_a = total stator current magnitude
- ω = electrical angular velocity

Additionally, simulation results confirm the smooth transition between base speed and extended-speed zones, validating the correct functioning of the flux weakening loop and its positive impact on vehicle performance.

An IPMSM forms the basis for the high-performance vector control strategy used for an electric vehicle (EV) drive train. The intention is to provide accurate torque production, efficient energy consumption, and robust dynamic performance across a wide range of operating conditions. The Fig [2] represents a cascade type SMC strategy consisting of a Field-Oriented Control (FOC) for low-speed performance and a Flux Weakening (FW) for high-speed performance was implemented to ensure the performance is achieved [11].



The diagram illustrates a speed feedback control system for a Permanent Magnet Synchronous Motor (PMSM). The control loop starts with a reference speed ω_{ref} , which is compared with the estimated speed ω from the Speed Estimator. The resulting speed error is processed by the Speed Controller to produce a reference d-axis current i_{sdref} . This reference current is then compared with the actual d-axis current i_{sd} and fed into the d-axis Current Controller. Similarly, the reference q-axis current i_{sqref} is compared with the actual q-axis current i_{sq} and fed into the q-axis Current Controller. The outputs of these controllers are the d-axis voltage u_{sd} and q-axis voltage u_{sq} , which are transformed by the Inverse Park Transformation into three-phase voltages u_{sa} , u_{sb} , and u_{sc} . These voltages are applied to the 3-phase PWM Voltage Inverter, which drives the PMSM. The motor's three-phase currents i_a , i_b , and i_c are measured and transformed by the Clarke Transformation into i_{sa} and i_{sb} . These are then transformed by the Park Transformation into the d-q axis currents i_{sd} and i_{sq} , which are fed back to the current controllers and the speed estimator. The speed estimator also receives the motor's mechanical angle θ to provide the estimated speed ω for the feedback loop.

Fig 3: block diagram of Field-Oriented Control method

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The battery voltage represented in Fig [6] of plot [3] was largely stable in the 320–330 V range, with brief voltage dips during heavy current draw and minor rises during regenerative events, confirming reliable battery behaviour under dynamic loading.

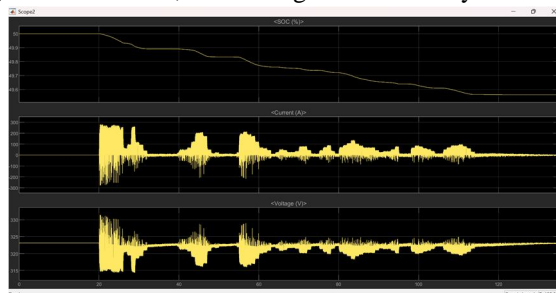


Fig 6: SOC, Current and Voltage Waveforms

The Fig [7] shows the mechanical load torque on the motor, extracted during the drive cycle, exhibited frequent oscillations in line with speed variations. These variations accurately replicated real-world vehicular forces such as rolling resistance, road slope, and inertia-induced torque, thereby validating the simulation model's realism and mechanical coupling accuracy.

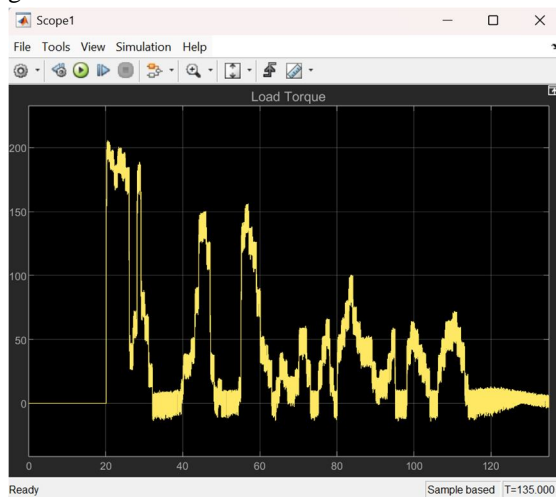


Fig 7: Load Torque waveform

The three-phase stator currents of the IPMSM in Fig [8] revealed a symmetrical sinusoidal profile modulated by high-frequency PWM switching from the inverter. Initial current magnitudes were higher during vehicle start-up, gradually stabilizing during steady-state operation. This reflects the FOC system's capability to manage varying current demands while maintaining waveform balance and motor protection.

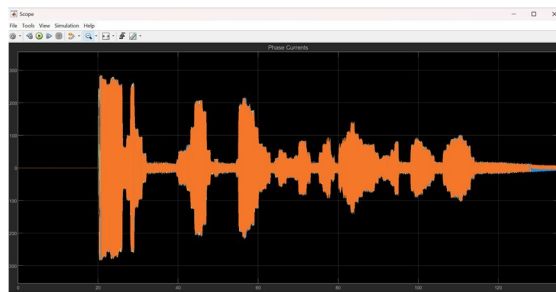


Fig 8: 3-phase current waveform

Additionally, a series of architectural simulations demonstrate the robustness of the complete EV control structure. The integrated driver model responded effectively to the FTP-75 cycle by continuously adjusting the acceleration and braking signals based on the speed error. These signals were converted into current references, which were tracked precisely by the FOC controller using Clarke and Park transformations.

The inverter was activated by the gate signals that resulted, providing the motor with the proper three-phase voltages. Dynamic torque and flux management, which are essential for effective propulsion, were preserved by this close control loop. The complete EV model also included four-wheel dynamics and environmental interactions, providing feedback to the control system for accurate vehicle behavior.

The overall system displayed stable operation throughout the simulation window, confirming that the coordinated implementation of FOC and FW with real-world drive cycles results in a well-balanced, energy-efficient, and high-performance electric vehicle propulsion system by the Figs [5], [6], [7] & [8].

VI. CONCLUSION

The efficiency of combining feedforward Flux Weakening (FW) and Field-Oriented Control (FOC) techniques over a broad operating speed range was shown by the simulation study and control implementation of the IPMSM-based electric vehicle propulsion system. The model, developed and executed in MATLAB/Simulink, was subjected to the standardized FTP-75 urban driving cycle, which closely mirrors real-world stop-and-go traffic conditions. The results affirm that the proposed control scheme ensures precise torque regulation at low speeds and extends motor capability beyond base speed without compromising system stability or energy efficiency. The vehicle exhibited responsive speed tracking, consistent torque output, and smooth transitions during acceleration and deceleration phases. The inverter voltage and three-phase current waveforms remained within safe and operational limits, demonstrating robust electrical performance. Battery behaviour, in terms of current, voltage, and SOC, confirmed efficient energy usage and successful implementation of regenerative braking during deceleration periods. The validity of the system was also demonstrated by realistic handling of the load torque fluctuations, which are demanded by the driving profile. Together, these results demonstrate the practicability and appropriateness of applying the novel IPMSM and advanced control strategies for EVs. The approach developed in this research work, beyond enhancing the drivability and the energy efficiency of the vehicle, offers to the EV system a rate of scalability capable to meet the efficiency and the performance requirements of the modern electric mobility solutions.

REFERENCES

- [1] A.Chandekar and R. T. Ugale, "Interior Permanent Magnet Synchronous Traction Motor for Electric Vehicle (EV) Application Over Wide Speed Range," 2022 IEEE International Conference on Power Electronics, Drives and Energy Systems (PEDES), Jaipur, India, 2022, pp. 1-6, doi: 10.1109/PEDES56012.2022.10080311.
- [2] Sheshadri Shekhar Rauth, Banshidhari Samanta. "Comparative analysis of IM/BLDC/PMSM drives for electric vehicle traction applications using ANN-based FOC." IEEE 17th India Council International Conference (INDICON) 2020 Volume: 17 [DOI: 10.1109/INDICON49873.2020.9342237]
- [3] Xiangdong Liu, Hao chen, Jing Zhao, Anouar Belahcen "Research on the performances and parameters of interior PMSM used for electric vehicles." IEEE Transactions on Industrial Electronics Volume 63
- [4] P. Pillay, R. Krishnan. "Modelling, Simulation, and Analysis of Permanent-Magnet Motor Drives, Part I: The permanent-magnet synchronous motor drive." IEEE transactions on industry applications, vol. 25
- [5] P. Pillay and R. Krishnan, "Modelling of permanent magnet motor drives," in IEEE Transactions on Industrial Electronics, vol. 35, no. 4, pp. 537-541, Nov. 1988, doi: 10.1109/41.9176.
- [6] Dongyun Lu and N. C. Kar, "A review of flux-weakening control in permanent magnet synchronous machines," 2010 IEEE Vehicle Power and Propulsion Conference, pp. 1-6, doi: 10.1109/VPPC.2010.5728986
- [7] J. Wang, J. Wu, C. Gan and Q. Sun, "Comparative study of flux weakening control methods for PMSM drive over wide speed range," 2016 19th International Conference on Electrical Machines and Systems (ICEMS), 2016, pp. 1-6
- [8] Ma, Guifang "MTPA and flux weakening control of permanent magnet synchronous motor for electric vehicle." International Journal of Mechatronics and Applied Mechanics. 2018. pp. 71-79.
- [9] J. Simanek, J. Novak, O. Cerny and R. Dolecek, "FOC and flux weakening for traction drive with permanent magnet synchronous motor," 2008 IEEE International Symposium on Industrial Electronics, 2008, pp. 753-758, doi: 10.1109/ISIE.2008.4677099.
- [10] Li, Muyang, "Flux-Weakening control for permanent-magnet synchronous motors based on Z-source inverters" (2014). Master's Theses (2009). Paper 284.
- [11] J. M. Kim, S.-K.Sul, "Speed control of interior permanent magnet synchronous Motor Drive for Flux weakening operation", IEEE Trans. On Industry Applications, vol. 33, pp. 43-48,1997.
- [12] Yuliang Wen, Hanfeng Zheng, Fang Yang, Xiaofan Zeng, A novel MTPA and flux weakening method of stator flux-oriented control of PMSM, Transportation Safety and Environment, Volume 3
- [13] Unni, A. S. Kumar, R. Manoj, S. Sunil and J. S. V C, "Design and Simulation of Test-bed for of Emulation Electric Vehicle Dynamics," 2021 16thInternational Conference on Ecological Vehicles and Renewable Energies (EVER), 2021, pp. 1-6, doi: 10.1109/EVER52347.2021.9456618.



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